# Performance Analysis of Raptor Codes under Maximum Likelihood Decoding

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Abstract—In this paper, we analyze the maximum likelihood decoding performance of Raptor codes with a systematic low-density generator-matrix code as the pre-code. By investigating the rank of the product of two random coefficient matrices, we derive upper and lower bounds on the decoding failure probability. The accuracy of our analysis is validated through simulations. Results of extensive Monte Carlo simulations demonstrate that for Raptor codes with different degree distributions and pre-codes, the bounds obtained in this paper are of high accuracy. The derived bounds can be used to design near-optimum Raptor codes with short and moderate lengths.

Index Terms—Raptor codes; asymptotic analysis; maximum likelihood (ML) decoding; decoding failure probability.

#### I. Introduction

Rateless codes have been increasingly used in many telecommunication systems [1], [2], [3], [4], including cellular networks and satellite communication systems. Recent work has shown that, by employing rateless codes, wireless transmission efficiency and reliability can be dramatically improved [5], [6].

Rateless codes are a class of *forward error correction* (FEC) codes with special properties, which were initially designed for the *binary erasure channel* (BEC). Compared with conventional FEC codes with a fixed code rate, rateless codes have a number of advantages. Firstly, similar as *low-density parity-check* (LDPC) codes, rateless codes can be implemented with

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far less complex encoding and decoding algorithms, which are attractive for implementation. Secondly, as suggested by the name, rateless codes are suitable for any code rate. They can automatically adapt to instantaneous channel states and do not require feedback channels [1], [3], [5]. This is because they can generate a potentially limitless stream of coded symbols, and all source symbols can be correctly decoded when there are a sufficient number of successfully received coded symbols. Hence, rateless codes are desirable for certain channels, such as erasure multicast or broadcast channels, whose real-time channel erasure probability is very difficult to capture or estimate. Furthermore, they have the potential to replace the conventional *automatic repeat request* (ARQ) mechanism as a new mechanism of transmission control protocol [7].

Among the well-known rateless codes, two codes stand out. One is the *Luby transform* (LT) codes [3], which are the first class of practical digital fountain codes with an average decoding cost in the order of  $O(k \log(k))$  where k is the number of source symbols. The other is the Raptor codes [1], which are the first class of fountain codes with linear time encoding and decoding complexities. Raptor codes are concatenated codes, which combine a traditional FEC code with an LT code to relax the condition that all input (source) symbols need to be recovered in an LT decoder. Note that Raptor codes have already been standardized in *the 3rd Generation Partnership Project* (3GPP) [4] to efficiently disseminate data over a broadcast/multicast network to provide multimedia broadcast and multicast services.

Despite the successful application of Raptor codes in 3GPP, our understanding of Raptor codes is still incomplete due to a lack of complete theoretical analysis on their decoding error performance. Without analytical results, the optimization of the degree distribution and other parameters of Raptor codes would be extremely difficult.

In this paper, we investigate the performance of Raptor codes by theoretically analyzing their decoding failure probability under *maximum likelihood* (ML) decoding. The decoding failure probability is the probability that not all source symbols can be decoded by ML decoding from a given number of successfully received coded symbols. We consider a Raptor code ensemble with a systematic  $(n, k, \eta)$  low-density generator matrix (LDGM) code as the pre-code. In the case of the erasure channel, ML decoding is equivalent to solving a consistent system of m linear equations in k unknowns

by means of *Gaussian elimination* (GE). In this paper, we investigate the decoding failure probability of Raptor codes by theoretically analyzing the rank of the product of two random coefficient matrices and deriving tight analytical bounds. The tightness of the bounds is confirmed by extensive Monte Carlo simulations. More specifically, the contributions of this paper are summarized in the following:

- Firstly, this paper provides analytical results (i.e. an upper bound and a lower bound) on the decoding failure performance of Raptor codes, using a systematic LDGM code as the pre-code and assuming ML decoding.
- Furthermore, simulations are conducted to validate the accuracy of the proposed bounds. That is, Raptor codes with different degree distributions and pre-codes are evaluated to verify the claims on the accuracy of the derived upper and lower bounds.

The rest of the paper is organized as follows. Section II reviews the related work. In Section III, a brief review of the encoding and decoding process of Raptor codes is given. In Section IV, a performance analysis of Raptor code is conducted by deriving an upper bound and a lower bound on the probability that not all source symbols can be successfully decoded by a receiver with a given number of successfully received coded symbols. Section V validates the analytical results through simulations, followed by concluding remarks in Section VI.

#### II. RELATED WORK

In this section, we review related work on the analysis of the performance of Raptor codes.

In general, there are two inter-related metrics to measure the performance of Raptor codes. One is the bit error probability and the other is the decoding failure probability. To analyze the bit error probability of Raptor codes, Rahnavard et al. [7] proposed a method to compute the upper and the lower bounds on the bit error probability of Raptor codes under ML decoding over the binary erasure channels (BEC). Despite the advances in [7], their work can be further improved in the following aspects. Firstly, the authors in [7] used a stochastic parity-check codes, i.e.  $(n, k, \eta)$  LDPC code, as the pre-code of Raptor codes. All entries of the parity check matrix are assumed to be independent and identically distributed (i.i.d) Bernoulli random variables [7]. Contrary to this assumption, in 3GPP standard [4], the pre-code of the standardized Raptor codes is a systematic LDGM code. The use of the systematic LDGM code as the pre-code is to guarantee that the parity check matrix is a full-rank matrix. Secondly, Rahnavard et al. assumed that the erasures on intermediate bit level are independent. As explained in [8, Ch. 6.2.1], this assumption would only hold if a very long interleaver was used. Using an interleaver in this setup, however, is not reasonable. In [9], the authors derived the upper and the lower bounds on the bit error probability of Raptor codes over Rayleigh fading channels assuming ML decoding.

In [1], Shokrollahi analyzed the decoding failure probability of Raptor codes with a finite length assuming *belief propagation* (BP) decoding. The analysis relies on the computation of

the failure probability of the LT codes under BP decoding, which was derived in [10]. ML decoding, on the other hand, is more computationally demanding than BP decoding for codes with a large length. The analysis of the decoding failure probability assuming ML decoding is however both important and significant, because it provides a benchmark on the optimum system performance that can be used to gauge the performance of other decoding schemes. Furthermore, in [8] a pseudo upper bound on the performance of Raptor codes under ML decoding was derived, under the assumption that the number of erasures correctable by the pre-code is small. This approximation is accurate only when the rate of the pre-code is sufficiently high. For the more general case, the decoding failure probability of Raptor codes still remains an open problem. In [11] it is shown that the rank profile of the constraint matrix of a Raptor code depends on the rank profile of the pre-code parity check matrix and the generator matrix of the LT code. The rank profile of the Raptor code cannot be determined if the rank profile of an LT code with a general degree distribution is unknown. In our previous work [6], we analyzed the rank profile of an LT code with a general degree distribution.

In this paper, we present theoretical analysis on the decoding failure probability of Raptor codes under ML decoding. We consider a Raptor code ensemble with a systematic  $(n,k,\eta)$  LDGM code as the pre-code to guarantee that the parity check matrix is a full-rank matrix. Furthermore, we take into account the fact that the residual erasure events after LT decoding are not independent, thereby deriving tighter bounds.

#### III. BACKGROUND OF RAPTOR CODES

This section is provided to familiarize the readers with the basic idea of Raptor codes, their encoding and decoding algorithms.

The encoding process of a Raptor code [1] is carried out in two phases: a) encode k source symbols with a (n, k) error correction code, which is referred to as the pre-code C, to form n intermediate symbols; b) encode the n intermediate symbols with an LT code. Each coded symbol is generated by the following encoding rules of LT codes [3]. Firstly, a positive integer d (often referred to as the "degree" of coded symbols) is drawn from the set of integers  $\{1, ..., n\}$  according to a probability distribution  $\Omega=(\Omega_1,...,\Omega_n)$ , where  $\Omega_d$  is the probability that d is selected and  $\sum_{d=1}^k \Omega_d=1$ . Then, ddistinct intermediate symbols are selected randomly and independently from the n intermediate symbols to form the coded symbol to be transmitted using the XOR operation, where each intermediate symbol is selected with equal probability. A Raptor code with parameters  $(k, \mathcal{C}, \Omega)$  is an LT code with distribution  $\Omega = (\Omega_1, ..., \Omega_n)$  on n symbols which are the output symbols of the pre-code C.

An illustration of a Raptor code is given in Fig. 1. In practice, the parity check matrix of the pre-code of Raptor codes is a deterministic matrix. For example, in 3GPP standard [4], the parity check matrix of the pre-code of the standardized Raptor codes is a systematic deterministic matrix. Using a systematic deterministic matrix as the pre-code ensures that the parity

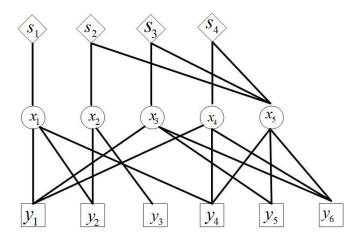


Figure 1. Two-stage structure of a Raptor code with a systematic pre-code.

check matrix of the pre-code is a full-rank matrix. However, it is difficult to obtain tractable analytical results of decoding performance for such Raptor codes. Therefore, in this paper we adopt a Raptor code ensemble with a semi-random  $(n,k,\eta)$  LDGM code as the pre-code for analytical tractability while ensuring that the parity check matrix of the pre-code is a full-rank matrix. The generator matrix of the pre-code, denoted by  $\mathbf{G}_{n\times k}^{\mathrm{pre}}$ , can be written as  $\mathbf{G}_{n\times k}^{\mathrm{pre}} = [\mathbf{I}_k|\mathbf{P}_{k\times(n-k)}]^T$ , where  $\mathbf{I}_k$  is an identity matrix of size k, and  $\mathbf{P}_{k\times(n-k)}$  is a k by (n-k) matrix whose entries are i.i.d. Bernoulli random variables with parameter  $\eta$ . Such a code is denoted as an  $(n,k,\eta)$  LDGM code. Furthermore, we can obtain the parity check matrix of this LDGM code as  $\mathbf{H}_{(n-k)\times n} = [\mathbf{P}_{(n-k)\times k}|\mathbf{I}_{(n-k)}]_{(n-k)\times n}$ .

Let m,  $(m \ge k)$ , be the number of coded symbols that have already been successfully received by a receiver and  $\gamma = \frac{m}{k}$ ,  $(\gamma \geq 1)$  be the overhead of reception. When a coded symbol is received by a receiver, we use a  $1 \times k$  binary row vector  $\mathbf{g}_i^{\mathrm{LT}}\mathbf{G}^{\mathrm{pre}}$  to represent the coding information contained in the coded symbol, where  $\mathbf{G}^{\mathrm{LT}}$  is a  $k\gamma \times n$  binary matrix,  $\mathbf{g}_i^{\mathrm{LT}}$  is the  $i^{th}$  row vector of  $\mathbf{G}^{\mathrm{LT}}$  and  $\mathbf{G}^{\mathrm{pre}}$  is a  $n \times k$  binary matrix. Let  $[\mathbf{G}]_{i,j}$  be the entry in the  $i^{th}$  row and the  $j^{th}$  column of the matrix **G**. Particularly,  $\left[\mathbf{g}_{i}^{LT}\right]_{1,j}$  is 1 if the coded symbol is a result of the XOR operation on the  $j^{th}$  intermediate symbol (and other intermediate symbols); otherwise  $\left[\mathbf{g}_{i}^{LT}\right]_{1,j}$  equals 0. For  $[\mathbf{G}^{\text{pre}}]_{i,j}$ , it is 1 if the  $i^{th}$  intermediate symbol is a result of the XOR operation on the  $j^{th}$  source symbol (and other source symbols); otherwise  $[\mathbf{G}^{\text{pre}}]_{i,j}$  equals 0. Therefore, a random row vector in this paper refers to the row vector of a randomly chosen coded symbol where the coded symbol is generated using the Raptor encoding process described above. Recall that  $\mathbf{s} = (s_1, s_2, ..., s_k)$  represents the k source symbols to be transmitted. The coded symbol can be expressed as:  $y_i = \mathbf{g}_i^{\mathrm{LT}} \mathbf{G}^{\mathrm{pre}} \mathbf{s}^T$ , where " $\mathbf{s}^T$ " is the transpose of  $\mathbf{s}$ .

Raptor codes can be decoded using a variety of decoding algorithms. A commonly used decoding algorithm for Raptor codes is the so-called "LT process" [3], but it is well known that the LT process is unable to decode all source symbols which can be possibly recovered from the received coded symbols. For example, the LT process relies on the existence of at least one degree-one coded symbol to be received in order

to start the decoding process. For Raptor codes with limited lengths, ML decoding algorithm [12] has been proposed to replace the LT process. The performance of ML decoding is the same as the Gaussian elimination. One way to apply the Gaussian elimination on Raptor codes is to solve a system of linear equations given in the following.

$$\mathbf{G}_{k\gamma \times n}^{\mathrm{LT}} \mathbf{G}_{n \times k}^{\mathrm{pre}} \mathbf{s}_{k \times 1}^{T} = \mathbf{y}_{k\gamma \times 1},$$

where  $\mathbf{y}_{k\gamma\times 1}=(y_1,y_2,...,y_{k\gamma})^T$ . Then we can obtain the following Lemma.

**Lemma 1.** A receiver can recover all k source symbols from the  $k\gamma$  coded symbols under ML decoding if and only if  $(\mathbf{G}_{n \times k}^{LT})_{k \gamma \times k} = \mathbf{G}_{n \times k}^{pre})_{k \gamma \times k}$  is a full-rank matrix, i.e. its rank equals k [1].

Note that in this paper, all algebraic operations and the associated analysis are conducted in a binary field GF(2).

#### IV. PERFORMANCE ANALYSIS OF RAPTOR CODES

Denote by  $A_{k\gamma}^k$  the event that a receiver can successfully decode all k source symbols conditioned on the event that the receiver has successfully received  $k\gamma$  coded symbols. Obviously the event that  $(\mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}\mathbf{G}_{n\times k}^{\mathrm{pre}})_{k\gamma\times k}$  is a full-rank matrix is equivalent to the event  $A_{k\gamma}^k$ . Let  $\overline{A_{\gamma k}^k}$  be the complement of event  $A_{k\gamma}^k$ . The main results of this paper are summarized in Theorems 2 and 3.

In this section, we shall analyze the probability  $\Pr\left[\overline{A_{\gamma k}^k}\right]$ . The analysis of decoding failure probability  $P_{k,n,\gamma}^{DF}=\Pr\left[\overline{A_{\gamma k}^k}\right]$  is conducted by analyzing the probability that the rank of  $(\mathbf{G}_{k\gamma \times n}^{\mathrm{LT}}\mathbf{G}_{n\times k}^{\mathrm{pre}})_{k\gamma \times k}$  is not k.

# A. Upper Bound on the Decoding Failure Probability of Raptor Codes

In this subsection, we will derive an upper bound on the decoding failure probability of Raptor codes with a systematic  $(n,k,\eta)$  LDGM code as the pre-code. The upper bound is formally stated in the following theorem.

**Theorem 2.** When a receiver successfully receives  $k\gamma$  coded symbols generated using the Raptor code  $(k, \mathcal{C}, \Omega(x))$ , where  $\mathcal{C}$  is an  $(n, k, \eta)$  LDGM code, and the coded symbols received at the receiver are decoded using ML decoding, the probability that not all k source symbols can be successfully decoded by a receiver with the  $k\gamma$ ,  $(k\gamma \geq k)$ , received coded symbols, denoted by  $P_{k,n,\gamma}^{DF}$ , is upper bounded by

$$P_{k,n,\gamma}^{DF} \leq \sum_{i=1}^{k} {k \choose i} \sum_{r=i}^{n-k+i} (J(r))^{k\gamma} D(i,r), \qquad (1)$$

where

$$J(r) = \sum_{d=1}^{n} \Omega_d \frac{\sum_{s=0,2,\dots,2 \lfloor \frac{d}{2} \rfloor} \binom{r}{s} \binom{n-r}{d-s}}{\binom{n}{d}}$$
 (2)

and

$$D(i,r) = {\binom{n-k}{r-i}} \left[ \frac{1 + (1-2\eta)^i}{2} \right]^{n-k-r+i}$$

$$\times \left[ \frac{1 - (1-2\eta)^i}{2} \right]^{r-i}$$
(3)

and  $\Omega_d$  is the degree distribution of LT codes.

Proof: See Appendix A.

### B. Lower Bound on the Decoding Failure Probability of Raptor Codes

In addition to the upper bound in the previous subsection, in the following paragraphs, we derive a lower bound on the decoding failure probability of Raptor codes which is formally stated in the following theorem.

**Theorem 3.** When a receiver successfully receives  $k\gamma$  coded symbols generated using the Raptor code  $(k, C, \Omega(x))$ , where C is an  $(n, k, \eta)$  LDGM code, and the coded symbols received at the receiver are decoded using ML decoding, the probability that not all k source symbols can be successfully decoded by a receiver with the  $k\gamma$ ,  $(k\gamma \ge k)$ , received coded symbols, denoted by  $P_{k,n,\gamma}^{DF}$ , is lower bounded by:

$$P_{k,n,\gamma}^{DF} \geq \sum_{i=1}^{k} {k \choose i} \sum_{r=i}^{n-k+i} (J(r))^{k\gamma} D(i,r)$$

$$-\frac{1}{2} \sum_{i=1}^{k} {k \choose i} \sum_{w_0=0}^{i} \sum_{w_1=i-w_0} \sum_{w_2=0}^{k-i} \mathbf{1}(w_0+w_2) \mathbf{1}(w_1+w_2)$$

$$\times {i \choose w_0} {k-i \choose w_2} \left\{ \sum_{r_0=w_0}^{n-k+w_0} \sum_{r_1=w_1}^{n-k+w_1} \sum_{r_0=w_2}^{n-k+w_2} D(w_0,r_0) D(w_1,r_1) \right\}$$

$$\times D(w_2,r_2) [J(r_0)J(r_1)J(r_2) + \overline{J}(r_0)\overline{J}(r_1)\overline{J}(r_2)] \}^{k\gamma}, \quad (4)$$

where  $\mathbf{1}(x)$  is an indicator function,  $\mathbf{1}(x) = 0$  if x = 0 and  $\mathbf{1}(x) = 1$  otherwise,  $\overline{J}(\cdot) = 1 - J(\cdot)$ ,  $D(w_0, r_0)$  is defined in Eq. (3) and  $J(r_0)$  is defined in Eq. (2).

#### C. A Special Case of the Derived Bounds

When we apply a special degree distribution - a binomial degree distribution [13] with  $\Omega_d = \frac{\binom{n}{d}}{(2^n-1)}, 1 \leq d \leq n$ , Eq. (1) can be further simplified into a much less (computationally) complex expression, for which Theorem 2 can be restated as the following Corollary.

**Corollary 4.** When a receiver successfully receives  $k\gamma$  coded symbols generated using the Raptor code  $(k, \mathcal{C}, \Omega(x))$  where  $\mathcal{C}$  is an  $(n, k, \eta)$  LDGM code,  $\Omega(x) = \sum_{d=1}^n \frac{\binom{n}{d} x^d}{(2^n-1)}$ , and the coded symbols received at the receiver are decoded using ML decoding, the probability that not all k source symbols can be successfully decoded by a receiver with the  $k\gamma$ ,  $(k\gamma \geq k)$ , received coded symbols, denoted by  $P_{k,n,\gamma}^{DF}$ , satisfies

$$P_{k,n,\gamma}^{DF} \le (2^k - 1)(\frac{(2^{n-1} - 1)}{(2^n - 1)})^{k\gamma}.$$
 (5)

Proof: See Appendix C.

Proof: See Appendix D.

For Theorem 3, we can simplify the lower bound into a less (computationally) complex expression as well. This is summarized in the following Corollary.

**Corollary 5.** When a receiver successfully receives  $k\gamma$  coded symbols generated using the Raptor code  $(k, \mathcal{C}, \Omega(x))$  where  $\mathcal{C}$  is an  $(n, k, \eta)$  LDGM code,  $\Omega(x) = \sum_{d=1}^n \frac{\binom{n}{d} x^d}{(2^n - 1)}$ , and the coded symbols received at the receiver are decoded using ML decoding, the probability that not all k source symbols can be successfully decoded by a receiver with the  $k\gamma$ ,  $(k\gamma \geq k)$ , received coded symbols, denoted by  $P_{k,n,\gamma}^{DF}$ , satisfies

$$P_{k,n,\gamma}^{DF} \ge (2^k - 1) \left[ \frac{(2^{n-1} - 1)}{(2^n - 1)} \right]^{k\gamma} - (2^k - 1)(2^{k-1} - 1)$$

$$\times \left\{ \left[ \frac{(2^{n-1} - 1)}{(2^n - 1)} \right]^3 + \left[ 1 - \frac{(2^{n-1} - 1)}{(2^n - 1)} \right]^3 \right\}^{k\gamma} . (6)$$

Compared with the general expressions in Theorems 2 and 3, the simplified expressions in Corollaries 4 and 5 allow us to easily observe the relationship between the decoding failure probability and the parameters of the encoding rules, i.e., k, n and  $\gamma$ . Additionally, the computation complexity of the derived upper bound can be reduced from  $O(\frac{1}{2}n^2k(n-k))$  to O(1).

upper bound can be reduced from  $O(\frac{1}{2}n^2k(n-k))$  to O(1). As for the lower bound, the computation complexity can be reduced from  $O(\frac{1}{8}n^6k^3(n-k)^3)$  to O(1).

#### V. SIMULATION RESULTS

In this section, we shall validate the accuracy of the analytical results and the tightness of the proposed bounds, using MATLAB simulations. Each point shown in the figures is the average result obtained from  $10^6$  simulations. For clarity, the simulation parameters adopted in this section are summarized in Table I.

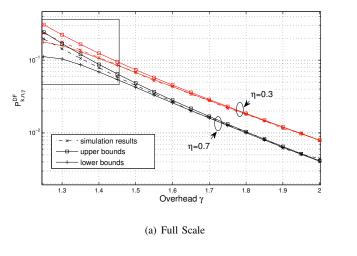
Table I SIMULATION PARAMETERS

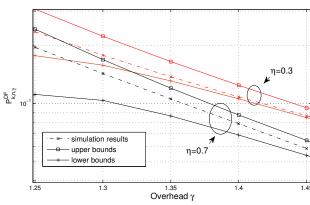
Rateless codes encoding parameters	
Number of source symbols k	20, 40, 70 and 100
Number of intermediate symbols $n$	21, 42, 73 and 105
Parameter for Bernoulli random variables $\eta$	0.3, 0.7
Pre-code $\mathcal C$	$(n, k, \eta)$ LDGM code
Degree distributions	
Standard degree distribution	$\Omega^{3GPP}(x)$
Binomial degree distribution	$\Omega_d = \frac{\binom{n}{d}}{(2^n - 1)}, 1 \le d \le n$ $\Omega_d = \frac{1}{d(d - 1)}, 2 \le d \le n$
Ideal soliton degree distribution	$\Omega_d = \frac{1}{d(d-1)}, 2 \leq d \leq n$
	and $\Omega_1 = \frac{1}{n}$

#### A. Verification of the Derived Bounds

In this subsection, the number of source symbols is set to be k=20 and the degree distribution of Raptor codes follows the widely used ideal soliton degree distribution [3]. Besides, the pre-code  $\mathcal C$  is assumed to be (21,20,0.3) and (21,20,0.7) LDGM codes respectively.

In Fig. 2(a) and 2(b), both analytical and simulation results are presented on  $P_{k,n,\gamma}^{DF}$ , the probability that not all k=20





(b) Zoom of the rectangular box in (a)

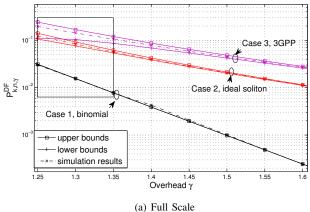
Figure 2. The decoding failure probabilities of Raptor codes with ideal soliton degree distribution and  $(n,k,\eta)$  LDGM codes as the pre-code versus overhead  $\gamma$ . Parameter for Bernoulli random variables  $\eta$  is set as 0.3 and 0.7.

source symbols can be successfully decoded by a receiver, for different values of the reception overhead  $\gamma=m/k$ . As shown in Fig. 2(a) and 2(b), our analytical results, i.e., the upper bound and the lower bound, match the simulation results very well. This validates the accuracy of the analysis. However, when the overhead  $\gamma$  is small, there is still a gap between the upper (lower) bound and simulation results in Fig. 2(a) and 2(b). The gap between the exact value and the upper bound is caused by the approximation used in Eq. (1), and the gap between the exact value and the lower bound is caused by Eq. (4).

### B. Investigation of the Impact of Degree Distribution on the Decoding Failure Probability

In this subsection, we investigate the performance for different distributions of LT codes when we fix the pre-code  $\mathcal{C}$  to be (21,20,0.7). The investigated degree distributions are divided into three cases.

- Case 1 uses the binomial degree distribution [13].
- Case 2 investigates the widely used ideal soliton degree distribution [3].



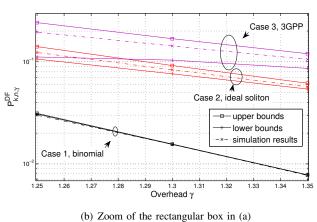


Figure 3. The decoding failure probabilities of Raptor codes with (n, k, 0.7) LDGM codes as the pre-code and different degree distributions versus overhead  $\gamma$ . The degree distributions of Raptor codes are chosen as ideal soliton degree distribution [3], the standardized degree distribution in 3GPP [4, Annex B] and binomial degree distribution [13].

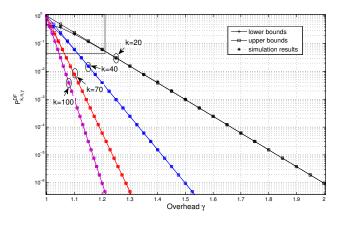
• Case 3 is the standardized degree distribution in 3GPP [4, Annex B]:

$$\begin{split} \Omega^{3GPP}(x) &= 0.0099x + 0.4663x^2 + 0.2144x^3 \\ &+ 0.1152x^4 + 0.1131x^{10} + 0.0811x^{11}. \end{split}$$

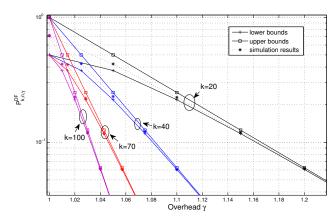
As shown in Fig. 3(a) and 3(b), for different degree distributions, our analytical bounds agree very well with the simulation results. The performance of Raptor codes with the binomial degree distribution outperforms those obtained with the other three degree distributions. Furthermore, the decoding failure probability of Raptor codes with the binomial degree distribution in Corollaries 4 and 5 are less computationally demanding compared with those in Theorems 2 and 3. Therefore, we will use Raptor codes with the binomial degree distribution in the following simulations.

### C. Investigation of the Impact of k on the Decoding Failure Probability of Raptor Codes

When the number of source symbols k varies from 20 to 100, our analytical results still match the simulation results very well. As shown in Fig. 4(a) and 4(b), at a larger value of the source symbols, a less reception overhead  $\gamma = m/k$ 



(a) Full Scale



(b) Zoom of the rectangular box in (a)

Figure 4. The decoding failure probabilities of Raptor codes with the binomial degree distribution and (n,k,0.7) LDGM codes as the pre-code at different values of the overhead  $\gamma$ . The number of source symbols k is set to be 20, 40, 70 and 100 respectively.

is required to achieve the same performance on the decoding failure probability.

#### VI. CONCLUSION

In this paper we studied the performance of finite-length Raptor codes with a systematic LDGM code as the precode, and derived an upper bound and a lower bound on the decoding failure probability of Raptor codes under ML decoding. Due to the concatenated coding structure of Raptor codes, we analyzed the rank behavior of the product of two random matrices to obtain the decoding failure probability. Furthermore, by considering a special degree distribution, i.e. the binomial degree distribution, we derived the simplified upper and lower bounds. On the basis of the results presented in the paper, we shall explore the optimum degree distribution and optimal parameter setting of Raptor codes in different channels as our future work.

#### ACKNOWLEDGMENT

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#### APPENDIX A

#### PROOF OF THEOREM 2

In this appendix, we prove Theorem 2.

According to the property of the matrix product [14, Eq. (4.5.1)], we have

$$\begin{aligned} & rank(\mathbf{G}_{n\times n}^{\mathrm{LT}}\mathbf{G}_{n\times k}^{\mathrm{pre}}) \\ &= & rank(\mathbf{G}_{n\times k}^{\mathrm{pre}}) - \dim\{N(\mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}) \cap R(\mathbf{G}_{n\times k}^{\mathrm{pre}})\}, \end{aligned} \tag{7}$$

where  $N(\bullet)$  is the right-hand null space of a matrix,  $R(\bullet)$  is the column vector space generated by a matrix and  $\dim\{\mathcal{V}\}$  represents the number of vectors in any basis for a vector space  $\mathcal{V}$ . It follows from the definition of  $\mathbf{G}_{n\times k}^{\mathrm{pre}}$  given earlier that the rank of  $\mathbf{G}_{n\times k}^{\mathrm{pre}}$  is k. It can then be readily obtained that

$$\begin{aligned} P_{k,n,\gamma}^{DF} &= \Pr[rank(\mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}\mathbf{G}_{n\times k}^{\mathrm{pre}}) \neq k] \\ &= \Pr[\dim\{N(\mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}) \cap R(\mathbf{G}_{n\times k}^{\mathrm{pre}})\} \neq 0]. \end{aligned} \tag{8}$$

For convenience, let  $W_{k\gamma,n,k}$  be the event that  $\dim\{N(\mathbf{G}_{k\gamma\times n}^{\mathrm{LT}})\cap R(\mathbf{G}_{n\times k}^{\mathrm{pre}})\}\neq 0$ . Now we need to analyze  $P_{k,n,\gamma}^{\mathrm{DF}}=\Pr[W_{k\gamma,n,k}].$  Provided that  $\mathbf{G}_{n\times k}^{\mathrm{pre}}$  is the generator matrix of a systematic  $(n,k,\eta)$  LDGM code, the event  $\dim\{N(\mathbf{G}_{k\gamma\times n}^{\mathrm{LT}})\cap R(\mathbf{G}_{n\times k}^{\mathrm{pre}})\}\neq 0$ , denoted by  $W_{k\gamma,n,k}$ , is equivalent to the event that at least one column vector  $R(\mathbf{G}_{n\times k}^{\mathrm{pre}})$  is in  $N(\mathbf{G}_{k\gamma\times n}^{\mathrm{LT}})$ , i.e.,  $\bigcup_{\mathbf{x}\in R(\mathbf{G}_{n\times k}^{\mathrm{pre}})}\mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}\mathbf{x}=\mathbf{0}$ , where  $\mathbf{x}$  is a column vector of  $R(\mathbf{G}_{n\times k}^{\mathrm{pre}})$ . It can be readily shown that

$$\Pr[W_{k\gamma,n,k}] = \Pr\left[\bigcup_{\mathbf{x} \in R(\mathbf{G}_{n\times k}^{\text{pre}})} \mathbf{G}_{k\gamma\times n}^{\text{LT}} \mathbf{x} = \mathbf{0}\right]$$

$$\leq \sum_{\mathbf{x} \in R(\mathbf{G}_{n\times k}^{\text{pre}})} \Pr\left[\mathbf{G}_{k\gamma\times n}^{\text{LT}} \mathbf{x} = \mathbf{0}\right]. \quad (9)$$

The column vector space  $R(\mathbf{G}_{n \times k}^{\mathrm{pre}})$  is partitioned into k subspace  $(\mathcal{V}_1, \mathcal{V}_2, \dots, \mathcal{V}_k)$  and  $\mathcal{V}_i$  is the subspace that contains all the column vectors which are summation of i column vectors of  $\mathbf{G}_{n \times k}^{\mathrm{pre}}$ . We denote  $\hat{\mathbf{I}} \mathbf{S}_i$  as the set of indices of the column vectors in  $\mathcal{V}_i$  and there are  $\binom{k}{i}$  elements in  $\hat{\mathbf{I}} \mathbf{S}_i$ . Let  $\mathbf{x}_a^i$  be the  $a^{th}, a \in \hat{\mathbf{I}} \mathbf{S}_i$  column vector in  $\mathcal{V}_i$ . It can be shown that

$$\sum_{\mathbf{x} \in R(\mathbf{G}_{n \times k}^{\text{pre}})} \Pr[\mathbf{G}_{k \gamma \times n}^{\text{LT}} \mathbf{x} = \mathbf{0}] = \sum_{i=1}^{k} \sum_{a \in \hat{\mathbf{i}} \S_{i}} \Pr[\mathbf{G}_{k \gamma \times n}^{\text{LT}} \mathbf{x}_{a}^{i} = \mathbf{0}].(10)$$

Observe that  $\mathbf{x}_a^i = \mathbf{G}_{n \times i}^a \mathbf{1}_i$  where  $\mathbf{G}_{n \times i}^a$  is the matrix formed by i column vectors selected from k column vectors of  $\mathbf{G}_{n \times k}^{\mathrm{pre}}$  and  $\mathbf{1}_i$  represent a  $i \times 1$  all one column vector. Let  $\left|\mathbf{x}_a^i\right|$  be the weight of column vector  $\mathbf{x}_a^i$ , using the law of total probability, we have

$$\Pr[\mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}\mathbf{x}_{a}^{i} = \mathbf{0}]$$

$$= \sum_{r=0}^{n} \Pr\left[\mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}\mathbf{x}_{a}^{i} = \mathbf{0} \middle| |\mathbf{x}_{a}^{i}| = r\right] \Pr\left[\left|\mathbf{x}_{a}^{i}\right| = r\right]. (11)$$

Firstly, we need to calculate  $\Pr\left[\left|\mathbf{x}_a^i\right|=r\right]$ . Provided  $\mathbf{G}_{n\times k}^{\mathrm{pre}}=\left[\mathbf{I}_k|\mathbf{P}_{k\times(n-k)}\right]^T$ , in the first k entries of  $\mathbf{G}_{n\times i}^a\mathbf{1}_i$  there are i

ones. If  $\left|\mathbf{x}_{a}^{i}\right|=r$ , then there are r-i ones in the last n-k entries of  $\mathbf{G}_{n\times i}^{a}\mathbf{1}_{i}$ , i.e.,  $\mathbf{P}_{(n-k)\times i}^{a}\mathbf{1}_{i}$ . Hence we can obtain that

$$\Pr\left[\left|\mathbf{x}_{a}^{i}\right|=r\right] = \Pr\left[\left|\mathbf{P}_{(n-k)\times i}^{a}\mathbf{1}_{i}\right|=(r-i)\right], (12)$$

and  $i \leq r \leq n-k+i$ . The rows of  $\mathbf{P}^a_{(n-k)\times i}$ , i.e.,  $\mathbf{p}_j, 1 \leq j \leq (n-k)$ , are random binary row vectors, which are generated independently. Each entry of  $\mathbf{P}^a_{(n-k)\times i}$  is i.i.d. Bernoulli random variable with parameter  $\eta$ . Therefore,  $\Pr[\mathbf{p}_j \mathbf{1}_i = 0] = \Pr[\mathbf{p}_{k,k\neq j} \mathbf{1}_i = 0]$ . The event that the  $j^{th}$  entry in  $\mathbf{x}^i_a$  is zero is equivalent to the event that there are even number of ones in row vector  $\mathbf{p}_j$ . Thus we have

$$\Pr[\mathbf{p}_{j}\mathbf{1}_{i} = 0] = \Pr[|\mathbf{p}_{j}| \text{ is even }]$$

$$= \sum_{s=0,2,...,2 \left\lfloor \frac{i}{2} \right\rfloor} {\binom{i}{s}} \eta^{s} (1-\eta)^{(i-s)}$$

$$= \frac{\left[(\eta + (1-\eta))^{i} + (-\eta + (1-\eta))^{i}\right]}{2}$$

$$= \frac{1 + (1-2\eta)^{i}}{2}.$$
(13)

There are  $\binom{n-k}{r-i}$  possible *combinations* for r-i ones in the last n-k entries. It follows that

$$\Pr\left[\left|\mathbf{P}_{(n-k)\times i}^{a}\mathbf{1}_{i}\right| = (r-i)\right]$$

$$= \binom{n-k}{r-i} \left\{\Pr\left[\mathbf{p}_{j}\mathbf{1}_{i} = 0\right]\right\}^{n-k-r+i}$$

$$\times \left\{1 - \Pr\left[\mathbf{p}_{j}\mathbf{1}_{i} = 0\right]\right\}^{r-i}.$$
(14)

Combining Eq. (12), (13) and (14), we can obtain that

$$D(i,r) = \Pr\left[\left|\mathbf{x}_{a}^{i}\right| = r\right]$$

$$= {n-k \choose r-i} \left[\frac{1 + (1-2\eta)^{i}}{2}\right]^{n-k-r+i}$$

$$\times \left[\frac{1 - (1-2\eta)^{i}}{2}\right]^{r-i}.$$
(15)

For  $\mathbf{x}_a^i, \mathbf{x}_{b,b \neq a}^i \in \mathcal{V}_i$ ,  $\mathbf{P}_{(n-k) \times i}^a$  and  $\mathbf{P}_{(n-k) \times i}^b$  have the same probability to form the same matrix formation. So we can obtain that  $\Pr\left[\left|\mathbf{P}_{(n-k) \times i}^a\mathbf{1}_i\right| = (r-i)\right] = \Pr\left[\left|\mathbf{P}_{(n-k) \times i}^b\mathbf{1}_i\right| = (r-i)\right]$ , which in turn leads to the conclusion that  $\Pr\left[\left|\mathbf{x}_a^i\right| = r\right] = \Pr\left[\left|\mathbf{x}_b^i\right| = r\right]$ . Now, we calculate  $\Pr\left[\mathbf{G}_{k\gamma \times n}^{LT}\mathbf{x}_a^i = 0 \mid \left|\mathbf{x}_a^i\right| = r\right]$ . The rows of  $\mathbf{G}_{\gamma k \times n}^{LT}$ , i.e.,  $\mathbf{g}_j^{LT}, 1 \leq j \leq k\gamma$ , are random binary row vectors, which are generated independently. We have

$$\Pr\left[\mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}\mathbf{x}_{a}^{i} = \mathbf{0} \middle| \left|\mathbf{x}_{a}^{i}\right| = r\right]$$
$$= \left\{\Pr\left[\mathbf{g}_{j}^{\mathrm{LT}}\mathbf{x}_{a}^{i} = 0 \middle| \left|\mathbf{x}_{a}^{i}\right| = r\right]\right\}^{k\gamma}.$$
 (16)

The degree of  $\mathbf{g}_{j}^{\mathrm{LT}}$ , i.e. the number of non-zero elements of  $\mathbf{g}_{j}^{\mathrm{LT}}$ , is chosen according to the pre-defined degree distribution  $\mathbf{\Omega} = (\Omega_{1}, ..., \Omega_{n})$  and each non-zero element is then placed randomly and uniformly into  $\mathbf{g}_{j}^{\mathrm{LT}}$ . It can be readily obtain that

$$\Pr\left[\mathbf{g}_{j}^{\mathrm{LT}}\mathbf{x}_{a}^{i}=0\middle|\left|\mathbf{x}_{a}^{i}\right|=r\right]$$

$$=\sum_{d=1}^{n}\Omega_{d}\Pr\left[\mathbf{g}_{j}^{\mathrm{LT}}\mathbf{x}_{a}^{i}=0\middle|\left|\mathbf{x}_{a}^{i}\right|=r,\left|\mathbf{g}_{j}^{\mathrm{LT}}\right|=d\right]. (17)$$

Let  $\mathbf{r}_{j}^{i} = (\mathbf{g}_{j1}^{\mathrm{LT}}\mathbf{x}_{a1}^{i}, \mathbf{g}_{j2}^{\mathrm{LT}}\mathbf{x}_{a2}^{i}, ..., \mathbf{g}_{jn}^{\mathrm{LT}}\mathbf{x}_{an}^{i})$ , where  $\mathbf{g}_{jk}^{\mathrm{LT}}$  is  $\left[\mathbf{g}_{j}^{\mathrm{LT}}\right]_{1,k}$  and  $\mathbf{x}_{ak}^{i}$  is  $\left[\mathbf{x}_{a}^{i}\right]_{k,1}$ . Then, we can obtain that

$$\Pr\left[\mathbf{g}_{j}^{\mathsf{LT}}\mathbf{x}_{a}^{i} = 0 \middle| |\mathbf{x}_{a}^{i}| = r, |\mathbf{g}_{j}^{\mathsf{LT}}| = d\right]$$

$$= \Pr\left[\left|\mathbf{r}_{j}^{i}\right| \text{ is even } \middle| |\mathbf{x}_{a}^{i}| = r, |\mathbf{g}_{j}^{\mathsf{LT}}| = d\right]$$

$$= \frac{\sum_{s=0,2,\dots,2\left\lfloor \frac{d}{2}\right\rfloor} \binom{r}{s} \binom{n-r}{d-s}}{\binom{n}{d}}.$$
(18)

Combining Eq. (17) and (18), we can obtain that

$$J(r) = \Pr\left[\mathbf{g}_{j}^{\mathrm{LT}}\mathbf{x}_{a}^{i} = 0 \middle| |\mathbf{x}_{a}^{i}| = r\right]$$

$$= \sum_{d=1}^{n} \Omega_{d} \frac{\sum_{s=0,2,\dots,2 \left\lfloor \frac{d}{2} \right\rfloor} \binom{r}{s} \binom{n-r}{d-s}}{\binom{n}{d}}. \quad (19)$$

Incorporating Eq. (16) into (19), it can be established that

$$\Pr\left[\mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}\mathbf{x}_{a}^{i}=\mathbf{0}\,\middle|\,\left|\mathbf{x}_{a}^{i}\right|=r\right]=\left[J(r)\right]^{k\gamma}.$$
 (20)

We can obtain that  $\Pr[\mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}\mathbf{x}_a^i=\mathbf{0}\mid |\mathbf{x}_a^i|=r]$  is only determined by the weight of  $\mathbf{x}_a^i$  rather than which i column vectors is chosen from  $\mathbf{G}_{n\times k}^{\mathrm{pre}}$  to obtain the summation  $\mathbf{x}_a^i$ . So we can conclude that  $\Pr[\mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}\mathbf{x}_a^i=\mathbf{0}]=\Pr[\mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}\mathbf{x}_b^i=\mathbf{0}]$ . Recall that there are  $\binom{k}{i}$  indices in  $\mathbf{i}\mathbf{\hat{S}}_i$ . Combining Eq. (15), (20), (11) and Eq. (10), yields the following results

$$P_{k,n,\gamma}^{DF} = \Pr[W_{k\gamma,n,k}]$$

$$\leq \sum_{i=1}^{k} \sum_{a \in \hat{\mathbb{I}}\hat{\mathbb{S}}_{i}} \Pr\left[\mathbf{G}_{k\gamma\times n}^{LT} \mathbf{x}_{a}^{i} = \mathbf{0}\right]$$

$$= \sum_{i=1}^{k} {n \choose i} \sum_{r=i}^{n-k+i} {n-k \choose r-i} \left[ \sum_{d=1}^{n} \Omega_{d} \frac{\sum_{s=0,2,\dots,2 \left\lfloor \frac{d}{2} \right\rfloor} {r \choose s} {n-r \choose d-s}}{{n \choose d}} \right]^{k\gamma}$$

$$\times \left[ \frac{1 + (1-2\eta)^{i}}{2} \right]^{n-k-r+i} \left[ \frac{1 - (1-2\eta)^{i}}{2} \right]^{r-i}, \quad (21)$$

which proves the theorem.

## APPENDIX B PROOF OF THEOREM 3

Similar as that in [7, Lemma 10], by using the Bonferroni inequality [15], we can obtain a lower bound of  $\Pr[W_{k\gamma,n,k}]$ 

$$P_{k,n,\gamma}^{DF} = \Pr[W_{k\gamma,n,k}]$$

$$= \Pr[\bigcup_{\mathbf{x} \in R(\mathbf{G}_{n\times k}^{\text{pre}})} \mathbf{G}_{k\gamma\times n}^{\text{LT}} \mathbf{x} = \mathbf{0}]$$

$$\stackrel{(a)}{\geq} \sum_{\mathbf{x} \in R(\mathbf{G}_{n\times k}^{\text{pre}})} \Pr[\mathbf{G}_{k\gamma\times n}^{\text{LT}} \mathbf{x} = \mathbf{0}]$$

$$-\frac{1}{2} \sum_{\mathbf{x}, \mathbf{y} \in R(\mathbf{G}_{n\times k}^{\text{pre}}), \mathbf{x} \neq \mathbf{y}} \Pr[\mathbf{G}_{k\gamma\times n}^{\text{LT}} \mathbf{x} = \mathbf{0} \cap \mathbf{G}_{k\gamma\times n}^{\text{LT}} \mathbf{y} = \mathbf{0}], (22)$$

where  $\mathbf{x} = \mathbf{G}_{n \times k}^{\text{pre}} \mathbf{a}, \mathbf{a} \in GF(2)^k$  and  $\mathbf{y} = \mathbf{G}_{n \times k}^{\text{pre}} \mathbf{b}, \mathbf{b} \in GF(2)^k \setminus \mathbf{a}$ . The first term can be calculated by using Theorem 2. Recall that  $\mathcal{V}_i$  is a subspace that contain all the column vectors which are summation of i column vectors of  $\mathbf{G}_{n \times k}^{\text{pre}}$ ,  $\mathbf{\hat{I}}\mathbf{\hat{S}}_i$  is the set of indices of the column vectors in  $\mathcal{V}_i$  and  $\mathbf{x}_i^a$ 

represents the  $a^{th}, a \in \mathbf{\hat{I}}\mathbf{\hat{S}}_i$  column vectors in  $\mathcal{V}_i$ . It can be readily shown that

$$\sum_{\mathbf{x},\mathbf{y}\in R(\mathbf{G}_{n\times k}^{\text{pre}}),\mathbf{x}\neq\mathbf{y}} \Pr[\mathbf{G}_{k\gamma\times n}^{\text{LT}}\mathbf{x} = \mathbf{0} \cap \mathbf{G}_{k\gamma\times n}^{\text{LT}}\mathbf{y} = \mathbf{0}]$$

$$= \sum_{\mathbf{x}\in R(\mathbf{G}_{n\times k}^{\text{pre}})} \sum_{\mathbf{y}\in R(\mathbf{G}_{n\times k}^{\text{pre}})\setminus\mathbf{x}} \Pr[\mathbf{G}_{k\gamma\times n}^{\text{LT}}\mathbf{x} = \mathbf{0} \cap \mathbf{G}_{k\gamma\times n}^{\text{LT}}\mathbf{y} = \mathbf{0}]$$

$$= \sum_{i=1}^{k} \sum_{a\in \hat{\mathbb{N}}_{i}} \sum_{\mathbf{y}\in R(\mathbf{G}_{n\times k}^{\text{pre}})\setminus\mathbf{x}_{a}^{i}} \Pr[\mathbf{G}_{k\gamma\times n}^{\text{LT}}\mathbf{x}_{a}^{i} = \mathbf{0} \cap \mathbf{G}_{k\gamma\times n}^{\text{LT}}\mathbf{y} = \mathbf{0}] (23)$$

where  $\mathbf{x}_a^i = \mathbf{G}_{n \times k}^{\mathrm{pre}} \mathbf{a}$ ,  $|\mathbf{a}| = i$ . Recall that  $\mathbf{y} = \mathbf{G}_{n \times k}^{\mathrm{pre}} \mathbf{b}$ ,  $\mathbf{b} \in GF(2)^k$ . We define three binary vectors  $\mathbf{z}_0$ ,  $\mathbf{z}_1$ , and  $\mathbf{z}_2 \in GF(2)^k$  such that for  $t = 1, ..., k, \mathbf{z}_0(t) = 1$  if and only if  $\mathbf{a}(t) = 1$  and  $\mathbf{b}(t) = 1$ ,  $\mathbf{z}_1(t) = 1$  if and only if  $\mathbf{a}(t) = 1$  and  $\mathbf{b}(t) = 0$ , and  $\mathbf{z}_2(t) = 1$  if and only if  $\mathbf{a}(t) = 0$  and  $\mathbf{b}(t) = 1$ . Let  $w_0, w_1$  and  $w_2$  be the weights of vectors  $\mathbf{z}_0$ ,  $\mathbf{z}_1$ , and  $\mathbf{z}_2$ , respectively. For  $\mathbf{x}_a^i$ , we have  $\mathbf{z}_0 + \mathbf{z}_1 = \mathbf{a}$  and  $\mathbf{z}_0 + \mathbf{z}_2 = \mathbf{b}$ . Hence we can obtain

$$\Pr\left[\mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}\mathbf{x}_{a}^{i} = \mathbf{0} \cap \mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}\mathbf{y} = \mathbf{0}\right]$$

$$= \Pr\left[\mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}\mathbf{G}_{n\times k}^{\mathrm{re}}\mathbf{z}_{0} = \mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}\mathbf{G}_{n\times k}^{\mathrm{pre}}\mathbf{z}_{1}\right]$$

$$\cap \mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}\mathbf{G}_{n\times k}^{\mathrm{pre}}\mathbf{z}_{1} = \mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}\mathbf{G}_{n\times k}^{\mathrm{pre}}\mathbf{z}_{2}$$

$$\left||\mathbf{z}_{0}| = w_{0} \cap |\mathbf{z}_{1}| = w_{1} \cap |\mathbf{z}_{2}| = w_{2}\right|. \tag{24}$$

Let  $I_{\mathbf{z}} = \{i_{\mathbf{z}1}, i_{\mathbf{z}2}, ..., i_{\mathbf{z}\tau}\}$  be the set of indices such that  $t \in I_{\mathbf{z}}$  for  $\mathbf{z}(t) = 1$ , we can obtain the sets of indices of vectors  $\mathbf{z}_0, \mathbf{z}_1$ , and  $\mathbf{z}_2$  as  $I_{\mathbf{z}_0}, I_{\mathbf{z}_1}$  and  $I_{\mathbf{z}_2}$ . Corresponding to the three sets  $I_{\mathbf{z}_0}, I_{\mathbf{z}_1}$  and  $I_{\mathbf{z}_2}$ , each column of the matrix  $\mathbf{G}^{pre}_{n \times k}$ ,  $\mathbf{g}^{pre}_i, 1 \leq i \leq k$ , can be divided into four mutually exclusive parts,  $\mathbf{g}_{\mathbf{z}_0}, \mathbf{g}_{\mathbf{z}_1}, \mathbf{g}_{\mathbf{z}_2}$  and  $\bigcup_{1 \leq i \leq k} \mathbf{g}^{pre}_i \setminus (\mathbf{g}_{\mathbf{z}_0} \cup \mathbf{g}_{\mathbf{z}_1} \cup \mathbf{g}_{\mathbf{z}_2})$ , i.e.,  $\mathbf{g}_{\mathbf{z}_0} \cap \mathbf{g}_{\mathbf{z}_1} = \{0\}$ . Let  $\mathbf{g}_{\mathbf{z}_0}$  be the subset of  $\bigcup_{1 \leq i \leq k} \mathbf{g}^{pre}_i$  such that all the elements of this subset are selected from  $\bigcup_{1 \leq i \leq k} \mathbf{g}^{pre}_i$  according to the indices in set  $I_{\mathbf{z}_0}$  and  $\mathbf{G}^{pre}_{\mathbf{z}_0}$  be the matrix whose columns are elements of  $\mathbf{g}_{\mathbf{z}_0}$ . The length of  $\mathbf{g}_{\mathbf{z}_0}$  is  $w_0$ . The same operation is applied to the formation of  $\mathbf{g}_{\mathbf{z}_1}$  and  $\mathbf{g}_{\mathbf{z}_2}$ , in which the elements are selected according to the indices in set  $I_{\mathbf{z}_1}$  and  $I_{\mathbf{z}_2}$ , and have lengths  $w_1$  and  $w_2$ , respectively. Let  $\mathbf{x}^{w_0} = \mathbf{G}^{pre}_{\mathbf{z}_0} \mathbf{1}_{w_0}$ ,  $\mathbf{x}^{w_1} = \mathbf{G}^{pre}_{\mathbf{z}_1} \mathbf{1}_{w_1}$  and  $\mathbf{x}^{w_2} = \mathbf{G}^{pre}_{\mathbf{z}_2} \mathbf{1}_{w_2}$ . Equivalently, Eq. (30) can be rewritten as,

$$\Pr\left[\mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}\mathbf{G}_{n\times k}^{\mathrm{pre}}\mathbf{z}_{0} = \mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}\mathbf{G}_{n\times k}^{\mathrm{pre}}\mathbf{z}_{1}\right]$$

$$\cap \mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}\mathbf{G}_{n\times k}^{\mathrm{pre}}\mathbf{z}_{1} = \mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}\mathbf{G}_{n\times k}^{\mathrm{pre}}\mathbf{z}_{2}$$

$$\left||\mathbf{z}_{0}| = w_{0} \cap |\mathbf{z}_{1}| = w_{1} \cap |\mathbf{z}_{2}| = w_{2}\right|$$

$$= \Pr\left[\mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}\mathbf{x}^{w_{0}} = \mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}\mathbf{x}^{w_{1}}\right]$$

$$\cap \mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}\mathbf{x}^{w_{1}} = \mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}\mathbf{x}^{w_{2}}\right]. \tag{25}$$

Recall that the rows of  $\mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}$ , i.e.,  $\mathbf{g}_{j}^{\mathrm{LT}}, 1\leq j\leq k\gamma$ , are random binary row vectors, which are generated independently. We have

$$\Pr\left[\mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}\mathbf{x}^{w_0} = \mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}\mathbf{x}^{w_1}\right]$$

$$\cap \mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}\mathbf{x}^{w_1} = \mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}\mathbf{x}^{w_2}\right]$$

$$= \left\{\Pr\left[\mathbf{g}_j^{\mathrm{LT}}\mathbf{x}^{w_0} = \mathbf{g}_j^{\mathrm{LT}}\mathbf{x}^{w_1}\right]\right\}^{k\gamma}.$$

$$\cap \mathbf{g}_i^{\mathrm{LT}}\mathbf{x}^{w_1} = \mathbf{g}_i^{\mathrm{LT}}\mathbf{x}^{w_2}\right\}^{k\gamma}.$$
(26)

According to the law of total probability, we have

$$\Pr\left[\mathbf{g}_{j}^{LT}\mathbf{x}^{w_{0}} = \mathbf{g}_{j}^{LT}\mathbf{x}^{w_{1}} \right]$$

$$\cap \mathbf{g}_{j}^{LT}\mathbf{x}^{w_{1}} = \mathbf{g}_{j}^{LT}\mathbf{x}^{w_{2}}$$

$$= \sum_{r_{0}=w_{0}}^{n-k+w_{0}} \sum_{r_{1}=w_{1}}^{n-k+w_{2}} \Pr[|\mathbf{x}^{w_{0}}| = r_{0}]$$

$$\times \Pr[|\mathbf{x}^{w_{1}}| = r_{1}] \Pr[|\mathbf{x}^{w_{2}}| = r_{2}]$$

$$\times \Pr\left[\mathbf{g}_{j}^{LT}\mathbf{x}^{w_{0}} = \mathbf{g}_{j}^{LT}\mathbf{x}^{w_{1}}$$

$$\cap \mathbf{g}_{j}^{LT}\mathbf{x}^{w_{1}} = \mathbf{g}_{j}^{LT}\mathbf{x}^{w_{2}}$$

$$\left| |\mathbf{x}^{w_{0}}| = r_{0} \cap |\mathbf{x}^{w_{1}}| = r_{1} \cap |\mathbf{x}^{w_{2}}| = r_{2} \right]$$

$$(27)$$

For  $\Pr[|\mathbf{x}^{w_0}| = r_0]$ , this can be calculated by using Eq. (15). Because all algebraic operations are conducted in a binary field,  $\mathbf{g}_j^{\text{LT}}\mathbf{x}^{w_0}$  can only be 1 or 0. Eq. (27) can be further written as:

$$\Pr\left[\mathbf{g}_{j}^{LT}\mathbf{x}^{w_{0}} = \mathbf{g}_{j}^{LT}\mathbf{x}^{w_{1}} \cap \mathbf{g}_{j}^{LT}\mathbf{x}^{w_{1}} = \mathbf{g}_{j}^{LT}\mathbf{x}^{w_{2}} \right] \\ ||\mathbf{x}^{w_{0}}| = r_{0} \cap |\mathbf{x}^{w_{1}}| = r_{1} \cap |\mathbf{x}^{w_{2}}| = r_{2} \\ = \Pr\left[\mathbf{g}_{j}^{LT}\mathbf{x}^{w_{0}} = 0 \cap \mathbf{g}_{j}^{LT}\mathbf{x}^{w_{1}} = 0 \cap \mathbf{g}_{j}^{LT}\mathbf{x}^{w_{2}} = 0 \right] \\ ||\mathbf{x}^{w_{0}}| = r_{0} \cap |\mathbf{x}^{w_{1}}| = r_{1} \cap |\mathbf{x}^{w_{2}}| = r_{2} \\ + \Pr\left[\mathbf{g}_{j}^{LT}\mathbf{x}^{w_{0}} = 1 \cap \mathbf{g}_{j}^{LT}\mathbf{x}^{w_{1}} = 1 \cap \mathbf{g}_{j}^{LT}\mathbf{x}^{w_{2}} = 1 \right] \\ ||\mathbf{x}^{w_{0}}| = r_{0} \cap |\mathbf{x}^{w_{1}}| = r_{1} \cap |\mathbf{x}^{w_{2}}| = r_{2} \\ \end{bmatrix}. \tag{28}$$

Recall that  $\mathbf{x}^{w_0} = \mathbf{G}_{\mathbf{z}_0}^{\text{pre}} \mathbf{1}_{w_0}$ ,  $\mathbf{x}^{w_1} = \mathbf{G}_{\mathbf{z}_1}^{\text{pre}} \mathbf{1}_{w_1}$ ,  $\mathbf{x}^{w_2} = \mathbf{G}_{\mathbf{z}_2}^{\text{pre}} \mathbf{1}_{w_2}$  and the columns of  $\mathbf{G}_{\mathbf{z}_0}^{\text{pre}}$ ,  $\mathbf{G}_{\mathbf{z}_1}^{\text{pre}}$ ,  $\mathbf{G}_{\mathbf{z}_2}^{\text{pre}}$  are mutually exclusive to each other. So event that  $|\mathbf{x}^{w_0}| = r_0$  is independent of event that  $|\mathbf{x}^{w_1}| = r_1$  or  $|\mathbf{x}^{w_2}| = r_2$  and the event that  $\mathbf{g}_j^{\text{LT}} \mathbf{x}^{w_0} = 1$  is independent of event that  $\mathbf{g}_j^{\text{LT}} \mathbf{x}^{w_1} = 1$  or  $\mathbf{g}_j^{\text{LT}} \mathbf{x}^{w_2} = 1$ . Conditioned on  $|\mathbf{x}^{w_0}| = r_0, |\mathbf{x}^{w_1}| = r_1, |\mathbf{x}^{w_2}| = r_2$ , the first part in Eq. (28) can be expressed as:

$$\Pr\left[\mathbf{g}_{j}^{\mathrm{LT}}\mathbf{x}^{w_{0}} = 0 \cap \mathbf{g}_{j}^{\mathrm{LT}}\mathbf{x}^{w_{1}} = 0 \cap \mathbf{g}_{j}^{\mathrm{LT}}\mathbf{x}^{w_{2}} = 0\right]$$

$$\left||\mathbf{x}^{w_{0}}| = r_{0} \cap |\mathbf{x}^{w_{1}}| = r_{1} \cap |\mathbf{x}^{w_{2}}| = r_{2}\right]$$

$$= \Pr\left[\mathbf{g}_{j}^{\mathrm{LT}}\mathbf{x}^{w_{0}} = 0 \middle| |\mathbf{x}^{w_{0}}| = r_{0}\right]$$

$$\times \Pr\left[\mathbf{g}_{j}^{\mathrm{LT}}\mathbf{x}^{w_{1}} = 0 \middle| |\mathbf{x}^{w_{1}}| = r_{1}\right]$$

$$\times \Pr\left[\mathbf{g}_{j}^{\mathrm{LT}}\mathbf{x}^{w_{2}} = 0 \middle| |\mathbf{x}^{w_{2}}| = r_{2}\right]. \tag{29}$$

Based on the *previous* analysis, we know that  $\Pr[\mathbf{g}_j^{\mathrm{LT}}\mathbf{x}^{w_0} = 0 \ | \ | \mathbf{x}^{w_0}| = r_0]$  only relates to parameter  $r_0$ . Let  $D(w_0, r_0) = \Pr[\mathbf{x}^{\mathrm{LT}}\mathbf{x}^{w_0} = r_0]$  and  $J(r_0) = \Pr[\mathbf{g}_j^{\mathrm{LT}}\mathbf{x}^{w_0} = 0 \ | \ | \mathbf{x}^{w_0}| = r_0]$ . For  $J(r_0)$ , it can be calculated by using Eq. (17) and (18). Based on the previous analysis,, we know that  $J(r_0)$  only relates to parameter  $r_0$  and  $D(w_0, r_0)$  is affected by parameter  $r_0$  and  $w_0$ . Hence for the same parameters  $w_0$ ,  $w_1$  and  $w_2$ , Eq. (25) has the same result. Because  $\mathbf{x}_a^i \neq \mathbf{y}$ , we can obtain that  $w_1 + w_2 \neq 0$  and  $w_0 + w_2 \neq 0$ . For  $\mathbf{x}_a^i$ , when  $|\mathbf{z}_0| = w_0$ , we have  $w_1 = i - w_0$  and there are  $\binom{i}{w_0}$  possible combinations of  $\mathbf{z}_0$ . For  $\mathbf{z}_2$ , there are  $\binom{k-i}{w_2}$  possible combination of  $\mathbf{z}_2$  when  $|\mathbf{z}_2| = w_2$ . Inserting Eq. (25), (27), (28) and (29) into

(24), we can obtain:

$$\sum_{\mathbf{y} \in R(\mathbf{G}_{n \times k}^{\text{pre}}) \setminus \mathbf{x}_{a}^{i}} \Pr[\mathbf{G}_{k \gamma \times n}^{\text{LT}} \mathbf{x}_{a}^{i} = \mathbf{0} \cap \mathbf{G}_{k \gamma \times n}^{\text{LT}} \mathbf{y} = \mathbf{0}]$$

$$= \sum_{w_{0} = 0}^{i} \sum_{w_{1} = i - w_{0}} \sum_{w_{2} = 0}^{k - i} \mathbf{1}(w_{0} + w_{2}) \mathbf{1}(w_{1} + w_{2}) \binom{i}{w_{0}} \binom{k - i}{w_{2}}$$

$$\times \left\{ \sum_{r_{0} = w_{0}} \sum_{r_{1} = w_{1}} \sum_{r_{0} = w_{2}} D(w_{0}, r_{0}) D(w_{1}, r_{1}) D(w_{2}, r_{2}) \right\}$$

$$= \left[ J(r_{0}) J(r_{1}) J(r_{2}) + \overline{J}(r_{0}) \overline{J}(r_{1}) \overline{J}(r_{2}) \right] \right\}^{\gamma k}, \tag{36}$$

where 
$$\mathbf{1}(x) := \begin{cases} 0 & \text{if } x = 0 \\ 1 & \text{otherwise} \end{cases}$$
.

For  $\mathbf{x}_a^i, \mathbf{x}_{b,b\neq a}^i \in \mathcal{V}_i$ , the probability  $\sum_{\mathbf{x}_a^i \neq \mathbf{y}} \Pr\left[\mathbf{G}_{k\gamma \times n}^{\mathrm{LT}} \mathbf{x}_a^i = \mathbf{0} \cap \mathbf{G}_{k\gamma \times n}^{\mathrm{LT}} \mathbf{y} = \mathbf{0}\right]$  is affected by parameter i. So we can obtain that  $\sum_{\mathbf{x}_a^i \neq \mathbf{y}} \Pr[\mathbf{G}_{k\gamma \times n}^{\mathrm{LT}} \mathbf{x}_a^i = \mathbf{0} \cap \mathbf{G}_{k\gamma \times n}^{\mathrm{LT}} \mathbf{y} = \mathbf{0}] = \sum_{\mathbf{x}_b^i \neq \mathbf{y}} \Pr[\mathbf{G}_{k\gamma \times n}^{\mathrm{LT}} \mathbf{x}_a^i = \mathbf{0} \cap \mathbf{G}_{k\gamma \times n}^{\mathrm{LT}} \mathbf{y} = \mathbf{0}]$ . Recall that there are  $\binom{k}{i}$  indices in  $\mathbf{\hat{I}}\mathbf{\hat{y}}^i$ . We can get that

$$\sum_{\mathbf{x},\mathbf{y}\in R(\mathbf{G}_{n\times k}^{\mathrm{pre}}),\mathbf{x}\neq\mathbf{y}} \Pr[\mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}\mathbf{x} = \mathbf{0} \cap \mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}\mathbf{y} = \mathbf{0}]$$

$$= \sum_{i=1}^{k} \sum_{a\in \hat{\mathbf{i}}\S_{i}} \sum_{\mathbf{y}\in R(\mathbf{G}_{n\times k}^{\mathrm{pre}})\backslash \mathbf{x}_{a}^{i}} \Pr[\mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}\mathbf{x}_{a}^{i} = \mathbf{0} \cap \mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}\mathbf{y} = \mathbf{0}]$$

$$= \sum_{i=1}^{k} \binom{k}{i} \sum_{w_{0}=0}^{i} \sum_{w_{1}=i-w_{0}} \sum_{w_{2}=0}^{k-i} \mathbf{1}(w_{0}+w_{2})\mathbf{1}(w_{1}+w_{2})$$

$$\times \binom{i}{w_{0}}\binom{k-i}{w_{2}} \{\sum_{r_{0}=w_{0}}^{n-k+w_{0}} \sum_{r_{1}=w_{1}}^{n-k+w_{1}} \sum_{r_{0}=w_{2}}^{n-k+w_{2}} D(w_{0},r_{0})D(w_{1},r_{1})$$

$$\times D(w_{2},r_{2})[J(r_{0})J(r_{1})J(r_{2}) + \overline{J}(r_{0})\overline{J}(r_{1})\overline{J}(r_{2})]\}^{k\gamma}. \quad (31)$$

This completes the proof of Theorem 3.

#### APPENDIX C

#### PROOF OF COROLLARY 4

When the binomial degree distribution (the expurgated standard random ensemble) [8], [13], i.e.,  $\Omega_d = \frac{\binom{r}{d}}{(2^n-1)}, 1 \leq d \leq n$ , is inserted into Eq. (17), we can obtain that

$$\Pr[\mathbf{g}_{j}^{\text{LT}}\mathbf{x}_{a}^{i} = 0 \mid |\mathbf{x}_{a}^{i}| = r]$$

$$= (2^{n} - 1)^{-1} \sum_{d=1}^{n} \sum_{s=0,2,\dots,2 \left\lfloor \frac{d}{2} \right\rfloor} {r \choose s} {n-r \choose d-s}.$$
 (32)

Similar to [13, Lemma 2], when the upper limit of the inner summation is changed from  $2\left\lfloor\frac{d}{2}\right\rfloor$  to  $2\left\lfloor\frac{n}{2}\right\rfloor$ , it will not affect the result of Eq. (32). This is because  $\binom{n-r}{d-s}$  with  $s>2\left\lfloor\frac{d}{2}\right\rfloor$  equals 0.

$$\Pr[\mathbf{g}_{j}^{LT}\mathbf{x}_{a}^{i} = 0 \mid |\mathbf{x}_{a}^{i}| = r]$$

$$= (2^{n} - 1)^{-1} \sum_{d=1}^{n} \sum_{s=0,2,\dots,2 \left\lfloor \frac{n}{2} \right\rfloor} {r \choose s} {n-r \choose d-s}$$

$$\stackrel{a}{=} (2^{n} - 1)^{-1} \sum_{s=0,2,\dots,2 \left\lfloor \frac{n}{2} \right\rfloor} {r \choose s} \sum_{d=1}^{n} {n-r \choose d-s}. \quad (33)$$

The reason why the order of the two summations in Eq. (33) can be exchanged is because the inner summation variable s is now independent of the outer summation variable d. Note that  $1 \le d \le n$ . Now we want to change the lower limit of the inner summation of Eq. (33) from 1 to 0 without affecting its result.

$$\Pr\left[\mathbf{g}_{j}^{\text{LT}}\mathbf{x}_{a}^{i} = 0 \mid |\mathbf{x}_{a}^{i}| = r\right]$$

$$= (2^{n} - 1)^{-1} \left\{ \sum_{s=0,2,\dots,2 \left\lfloor \frac{n}{2} \right\rfloor} {r \choose s} \left[ \sum_{d=0}^{n} {n-r \choose d-s} - {n-r \choose d-s}_{d=0} \right] \right\}$$

$$\stackrel{b}{=} (2^{n} - 1)^{-1} \left[ \sum_{s=0,2,\dots,2 \left\lfloor \frac{n}{2} \right\rfloor} {r \choose s} \sum_{d=0}^{n} {n-r \choose d-s} - {r \choose s} {n-r \choose d-s}_{s=d=0} \right]. (34)$$

Step (b) is because the term  $\binom{n-r}{d-s}_{d=0}$  equals 0 for  $s \neq 0$ . Hence, only the case s=0 needs to be considered. The terms  $\binom{n-r}{d-s}$  restricts d to  $s \leq d \leq n-r+s$ , such that

$$\sum_{d=0}^{n} \binom{n-r}{d-s} = \sum_{d=s}^{n-r+s} \binom{n-r}{d-s} = \sum_{d=0}^{n-r} \binom{n-r}{d} = 2^{n-r}.$$
 (35)

Combining this term with the last expression for  $\Pr[\mathbf{g}_j^{LT}\mathbf{x}_a^i = 0 \mid |\mathbf{x}_a^i| = r]$  yields

$$\begin{aligned}
& \left[ \mathbf{g}_{j}^{\mathrm{LT}} \mathbf{x}_{a}^{i} = 0 \mid \left| \mathbf{x}_{a}^{i} \right| = r \right] \\
&= (2^{n} - 1)^{-1} \left( 2^{n-r} \sum_{s=0,2,\dots,2 \left\lfloor \frac{n}{2} \right\rfloor} {r \choose s} - 1 \right) \\
&= (2^{n} - 1)^{-1} (2^{n-r} 2^{r-1} - 1) \\
&= \frac{(2^{n-1} - 1)}{(2^{n} - 1)}, \quad (36)
\end{aligned}$$

where we have used identity  $\sum_{s \text{ even}} \binom{r}{s} = 2^{r-1}$ . We can observe that  $\Pr[\mathbf{g}_{j}^{\text{LT}}\mathbf{x}_{a}^{i} = 0 \mid |\mathbf{x}_{a}^{i}| = r]$  is independent from the weight of  $\mathbf{x}_{a}^{i}$ , hence  $\Pr[\mathbf{G}_{k\gamma\times n}^{\text{LT}}\mathbf{x}_{a}^{i} = 0 \mid |\mathbf{x}_{a}^{i}| = r] = \Pr[\mathbf{G}_{k\gamma\times n}^{\text{LT}}\mathbf{x}_{a}^{i} = 0]$ . Combining Eq. (16), (37), (10) and (8), we can obtain that

$$P_{k,n,\gamma}^{DF} = \Pr[W_{k\gamma,n,k}]$$

$$= \Pr\left[\bigcup_{\mathbf{x} \in R(\mathbf{G}_{n\times k}^{\text{pre}})} \mathbf{G}_{k\gamma\times n}^{\text{LT}} \mathbf{x} = 0\right]$$

$$\leq \sum_{\mathbf{x} \in R(\mathbf{G}_{n\times k}^{\text{pre}})} \Pr\left[\mathbf{G}_{k\gamma\times n}^{\text{LT}} \mathbf{x} = 0\right]$$

$$= (2^{k} - 1) \Pr\left[\mathbf{G}_{k\gamma\times n}^{\text{LT}} \mathbf{x} = 0 | |\mathbf{x}| = r\right]$$

$$= (2^{k} - 1) (\frac{(2^{n-1} - 1)}{(2^{n} - 1)})^{k\gamma}.$$
(38)

The proof of Corollary 4 is completed.

#### APPENDIX D

#### PROOF OF COROLLARY 5

When the binomial degree distribution is inserted into Eq. (13), by using the result of Eq. (37), we can obtain that

$$J(r_0) = \Pr[\mathbf{g}_j^{\text{LT}} \mathbf{x}^{w_0} = 0 | |\mathbf{x}^{w_0}| = r_0]$$

$$= \frac{(2^{n-1} - 1)}{(2^n - 1)}.$$
(39)

Insert Eq. (39) into Eq. (25), we can obtain that

$$\Pr\left[\mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}\mathbf{G}_{n\times k}^{\mathrm{pre}}\mathbf{z}_{0} = \mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}\mathbf{G}_{n\times k}^{\mathrm{pre}}\mathbf{z}_{1}\right] 
\cap \mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}\mathbf{G}_{n\times k}^{\mathrm{pre}}\mathbf{z}_{1} = \mathbf{G}_{k\gamma\times n}^{\mathrm{LT}}\mathbf{G}_{n\times k}^{\mathrm{pre}}\mathbf{z}_{2} 
| |\mathbf{z}_{0}| = w_{0} \cap |\mathbf{z}_{1}| = w_{1} \cap |\mathbf{z}_{2}| = w_{2}| 
= \sum_{r_{0}=w_{0}}^{n-k+w_{0}}\sum_{r_{1}=w_{1}}^{n-k+w_{2}}\sum_{r_{0}=w_{2}}^{n-k+w_{2}}D(w_{0}, r_{0})D(w_{1}, r_{1})D(w_{2}, r_{2}) 
\times \left\{\left[\frac{(2^{n-1}-1)}{(2^{n}-1)}\right]^{3}+\left[1-\frac{(2^{n-1}-1)}{(2^{n}-1)}\right]^{3}\right\}^{k\gamma} 
= \left\{\left[\frac{(2^{n-1}-1)}{(2^{n}-1)}\right]^{3}+\left[1-\frac{(2^{n-1}-1)}{(2^{n}-1)}\right]^{3}\right\}^{k\gamma}.$$
(40)

Incorporating Eq. (40) into Eq. (30), we can obtain that

$$\sum_{\mathbf{x}_{a}^{i} \neq \mathbf{y}} \Pr[\mathbf{G}_{k\gamma \times n}^{LT} \mathbf{x}_{a}^{i} = 0 \cap \mathbf{G}_{k\gamma \times n}^{LT} \mathbf{y} = 0]$$

$$= \sum_{w_{0}=0}^{i} \sum_{w_{1}=i-w_{0}} \sum_{w_{2}=0}^{k-i} \mathbf{1}(w_{0} + w_{2}) \mathbf{1}(w_{1} + w_{2}) \binom{i}{w_{0}} \binom{k-i}{w_{2}}$$

$$\times \left\{ \left[ \frac{(2^{n-1} - 1)}{(2^{n} - 1)} \right]^{3} + \left[ 1 - \frac{(2^{n-1} - 1)}{(2^{n} - 1)} \right]^{3} \right\}^{k\gamma}$$

$$= (2^{k} - 2) \left\{ \left[ \frac{(2^{n-1} - 1)}{(2^{n} - 1)} \right]^{3} + \left[ 1 - \frac{(2^{n-1} - 1)}{(2^{n} - 1)} \right]^{3} \right\}^{k\gamma}. \quad (41)$$

Combining Eq. (41), (23) and (22), we can obtain that

$$P_{k,n,\gamma}^{DF} = \Pr[W_{k\gamma,n,k}]$$

$$\geq \sum_{\mathbf{x} \in R(\mathbf{G}_{n\times k}^{\text{pre}})} \Pr[\mathbf{G}_{k\gamma\times n}^{\text{LT}} \mathbf{x} = 0]$$

$$-\frac{1}{2} \sum_{\mathbf{x}, \mathbf{y} \in R(\mathbf{G}_{n\times k}^{\text{pre}}), \mathbf{x} \neq \mathbf{y}} \Pr[\mathbf{G}_{k\gamma\times n}^{\text{LT}} \mathbf{x} = 0 \cap \mathbf{G}_{k\gamma\times n}^{\text{LT}} \mathbf{y} = 0]$$

$$= (2^{k} - 1) \left( \frac{(2^{n-1} - 1)}{(2^{n} - 1)} \right)^{k\gamma} - \frac{1}{2} \sum_{i=1}^{k} {k \choose i} (2^{k} - 2)$$

$$\times \left\{ \left[ \frac{(2^{n-1} - 1)}{(2^{n} - 1)} \right]^{3} + \left[ 1 - \frac{(2^{n-1} - 1)}{(2^{n} - 1)} \right]^{3} \right\}^{k\gamma}$$

$$= (2^{k} - 1) \left[ \frac{(2^{n-1} - 1)}{(2^{n} - 1)} \right]^{k\gamma} - (2^{k} - 1)(2^{k-1} - 1)$$

$$\times \left\{ \left[ \frac{(2^{n-1} - 1)}{(2^{n} - 1)} \right]^{3} + \left[ 1 - \frac{(2^{n-1} - 1)}{(2^{n} - 1)} \right]^{3} \right\}^{k\gamma}. \quad (42)$$

The proof of Corollary 5 is completed.

#### REFERENCES

- [1] A. Shokrollahi, "Raptor codes," *IEEE Trans. Inf. Theory*, vol. 52, no. 6, pp. 2551-2567, 2006.
- [2] J. W. Byers, M. Luby, M. Mitzenmacher, and A. Rege., "A digital fountain approach to reliable distribution of bulk data," SIGCOMM Comput. Commun. Rev., vol. 28, no. 4, pp. 56-67, 1998.
- [3] M. Luby, "LT codes," in *Proceedings of the 43rd IEEE FOCS*, pp. 271-280, 2002.
- [4] "3GPP TS 26.346 v6.1.0 (2005-06) Technical Specification Group Services and System Aspects; Multimedia Broadcast/Multicast Service; Protocols and Codecs," Tech. Rep.
- [5] H. D. T. Nguyen, T. Le-Nam, and H. Een-Kee, "On transmission efficiency for wireless broadcast using network coding and fountain codes," *IEEE Communications Letters*, vol. 15, no. 5, pp. 569-571, 2011.

- [6] P. Wang, G. Mao, Z. Lin, X. Ge, and B. Anderson, "Network coding based wireless broadcast with performance guarantee," *IEEE Trans. Wireless Communications*, vol. 14, no. 1, pp. 532-544, 2014.
- [7] N. Rahnavard, B. Vellambi, and F. Fekri, "Rateless codes with unequal error protection property," *IEEE Trans. Inf. Theory*, vol. 53, no. 4, pp. 1521-1532, April 2007.
- [8] B. Schotsch, "Rateless coding in the finite length regime," Ph.D. dissertation, Inst. of Commun. Systems and Data Proc., RWTH Aachen, Aachen, Germany, Jul 2014.
- [9] J. Yue, Z. Lin, B. Vucetic, G. Mao, and T. Aulin, "Performance analysis of distributed raptor codes in wireless sensor networks," *IEEE Trans. Communications*, vol. 61, no. 10, pp. 4357-4368, 2013.
- [10] R. Karp, M. Luby, and A. Shokrollahi, "Finite length analysis of LT codes," in *Proceedings of IEEE ISIT*, p. 39, 2004.
- [11] A. Shokrollahi and M. Luby, "Raptor codes," Found. and Trends on Commun. and Inf. Theory, vol. 6, pp. 213-322, Mar. 2009.
  [12] A. Shokrollahi, S. Lassen, and R. Karp, "Systems and processes for de-
- [12] A. Shokrollahi, S. Lassen, and R. Karp, "Systems and processes for decoding chain reaction codes through inactivation," US Patent 6,856,263.
  [Online]. Available: http://www.google.com/patents/US6856263, Feb. 15
  2005
- [13] B. Schotsch, H. Schepker, and P. Vary, "The performance of short random linear fountain codes under maximum likelihood decoding," in *Proceedings of IEEE ICC*, pp. 1-5, June 2011.
- [14] C. D. Meyer, Matrix Analysis and Applied Linear Algebra. Siam, 2000.
- [15] L. Comtet, Advanced Combinatorics. Reidel, 1974.



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